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Absolute Maximum Ratings

Recommended Operating Conditions

All voltages are referred to pin 1		(For operating circuits Figures 4 and 5)		
13, Supply Current (Figure 6)	45 mA	l5, Input Current During Sync Pulse	>5 µA	
15, Input Current	2 m A	VIN p-p, Composite Video Input Signal	3 (1 to 6)V	
V5, Input Voltage	-6V	110, Input Current During Line Flyback Pulse	0.2 to 2 mA	
12, Output Current	22 m A	Ig, Switch Over Current	>2 mA	
V2, Output Voltage	12V	td, Time Difference Between the Output Pulse		
18, Switch-Over Current for Video Recording	5 mA	at Pin 2 and the Line Flyback Pulse at 10	< 20 µs	
110, Flyback Peak Pulse Current	5 mA	13, Current Consumption (Figure 6)	≤45 m.A	
V11, Phase Correction Voltage	0 to V3	T _A , Ambient Operating Temperature Range	0°C to +60°C	
T _A , Ambient Temperature	60° C			

Electrical Characteristics $T_A = 25^{\circ}C$, $f_0 = 15,625$ Hz in the test circuit Figure 2 (Note 1)

SYMBOL	CHARACTERISTIC	CONDITIONS	MIN	TYP	MAX	UNITS
V7	Amplitude of the Frame Pulse			>8		v
t7	Frame Pulse Durations			>150		μs
ROUT 7	Output Resistance at Pin 7 (High State)	х. -	7.5	10	13	kΩ
t2	Output Pulse Duration	Typical Ranges	25		28	μs
V2 Res	Residual Output Voltage	l₂ = 20 mA		<0.55		V
fo	Oscillator Frequency	C13/1 = 10 nF, R14/1 = 10.5 k Ω	14063	15625	17187	Hz
±∆f F	Frequency Pull-In Range		400		1000	Hz
±∆fH	Frequency Holding Range	Typical Ranges	400		1000	Hz
df _o /dt _d	Slope of Phase Comparator Control Loop			2		kHz/µs
dtd/dtp	Gain of Phase Control			20		
tp	Phase Shift Between Leading Edge of Composite Video Signal and Line Flyback Pulse (Note 2) Adjustable by V11	Typical Range	0		3.5	μs

Note 1: By modification of the frequency-determining network at pins 13 and 14, these ICs can also be used for other line frequencies. Note 2: The limited flyback pulse should overlap the video signal sync pulse on both edges.

Functional Description

The sync separator separates the synchronizing pulses from the composite video signal. The noise inverter circuit, which needs no external components, in connection with an integrating and differentiating network frees the synchronizing signal from distortion and noise.

The frame sync pulse is obtained by multiple integration and limitation of the synchronizing signal, and is available at pin 7. The RC network, hitherto required between sync separator and frame oscillator is no longer needed. Since the frame sync pulse duration at pin 7 is subject to production spreads, it is recommended to use the leading edge of this pulse for triggering.

The frequency of the line oscillator is determined by a 10 nF polystyrene capacitor at pin 13 which is charged and discharged periodically by 2 internal current sources. The external resistor at pin 14 defines the charging current and consequently in conjunction with the oscillator capacitor the line frequency.

The phase comparator compares the sawtooth voltage of the oscillator with the line sync oulses. Simultaneously, an AFC voltage is generated which influences the oscillator frequency. A frequency range limiter restricts the frequency holding range.

The oscillator sawtooth voltage, which is in a fixed ratio to the line sync pulses, is compared with the flyback pulse in the phase control circuit, in this way compensating all drift of delay times in driver and line output stage. The correct phase position and hence the horizontal position of the picture can be adjusted by the 10 k Ω potentiometer connected to pin 11. Within the adjustable range the output pulse duration (pin 2) is constant. Any larger displacements of the picture, e.g., due to non-symmetrical picture tube, should not be corrected by the phase potentiometer, since in all cases the flyback pulse must overlap the sync pulse on both edges (Figure 3).

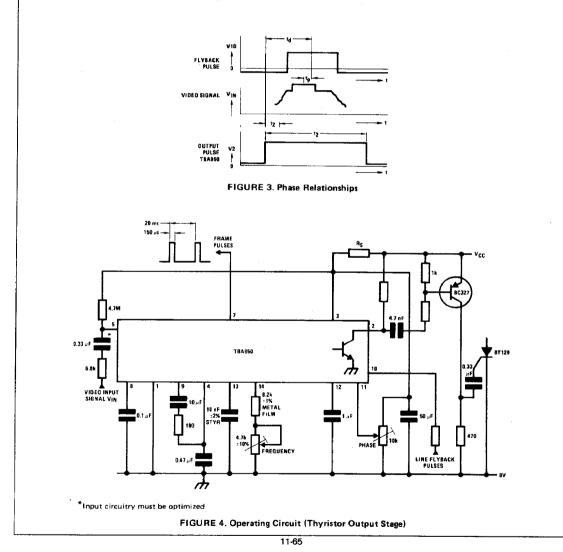
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Functional Description (Continued)

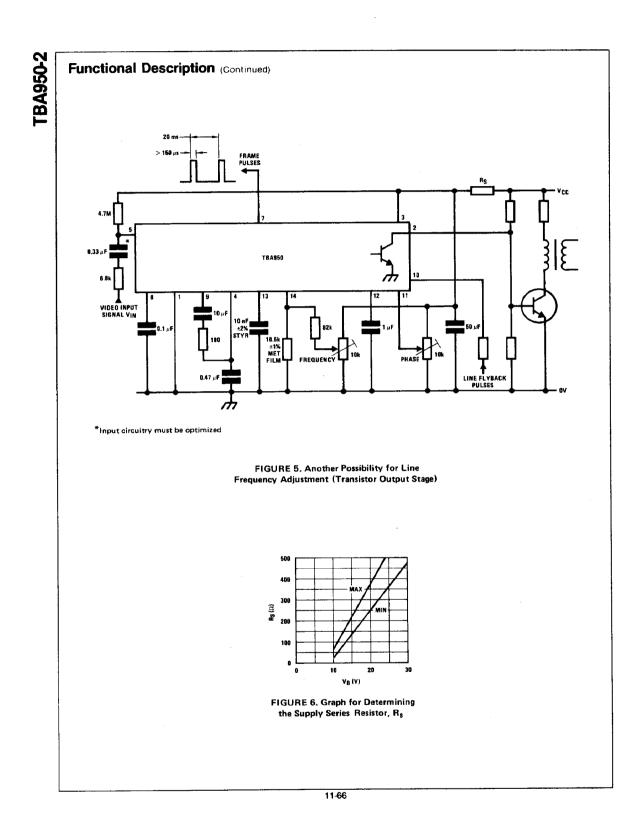
The switching stage has an auxiliary function. When the 2 signals supplied by the sync separator and the phase control circuit, respectively, are in synchronism, a saturated transistor is in parallel with the integrated 2 $k\Omega$ resistor at pin 9. Thus the time constant of the filter network at pin 4 increases and consequently reduces the pull-in range of the phase comparator circuit for the synchronized state to approximately 50 Hz. This arrangement ensures disturbance-free operation.

For video recording operation, this automatic switchover can be blocked by a positive current fed into pin 8, e.g., via a resistor connected to pin 3. It may also be useful to connect a resistor of about 680 Ω or 1 k Ω between pin 9 and earth. The capacitor at pin 4 may be lowered, e.g., to 0.1 μ F. These alterations do not significantly influence the normal operation of the IC and thus do not need to be switched.

The output stage delivers at pin 2 output pulses of duration and polarity suitable for driving the line driver stage. If the supply voltage goes down (e.g., by switching off the mains) a built-in protection circuit ensures defined line frequency pulses down to V3 = 4V and shuts off when V3 falls below 4V, thus preventing pulses of undefined duration and frequency. Conversely, if the supply voltage rises, pulses defined in duration and frequency will appear at the output pin as soon as V3 reaches 4.5V. In the range between V3 = 4.5V and full supply the shape and frequency of the output pulses are practically constant.



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